# Lecture 2: Basics for linear estimation, the white box case 

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## 1. Introduction

We present and illustrate the roadmap for a linear parameter estimation problem, when the structure of the model is known ('white box case'). The Ordinary Least Square case is first considered to introduce all the useful tools. We focus then on optimal ways to implement the best estimation through the study of the sensitivity matrix and other matrices depending on it.

## 2. The roadmap for solving a linear parameter estimation problem : the Ordinary Least Square case

### 2.1. Generate data

Suppose that we have realised an experiment that provides $m$ measurements $y_{i}=y\left(t_{i}\right)$ for $i=1, \ldots, m$ at $m$ discrete values of time $t$ (the 'independent' variable). These measurements are the components of the vector $(m \times 1)$ of experimental measurements $\boldsymbol{y}=\left[y_{1} \ldots y_{i} \ldots y_{m}\right]^{\mathrm{t}}$.

Times of measurements are regularly spaced between $t_{\text {min }}$ and $t_{\text {max }}$ and are the components of the time vector $(m \times 1) \boldsymbol{t}=\left[t_{\text {min }} \ldots t_{i} \ldots t_{\text {max }}\right]^{t}$ with $t_{i}=t_{\text {min }}+(i-1) d t$, for $i=1, \ldots, m$. Let $\varepsilon_{i}$ the (unknown) error associated to the measurement $y_{i}(i=1, \ldots, m)$, then the measurement errors vector $(m \times 1)$ is $\varepsilon=\left[\varepsilon_{1} \ldots \varepsilon_{i} \ldots \varepsilon_{m}\right]^{t}$. Some assumptions have to be done on these measurement errors. They are detailed in Table 1 (see also Beck et al., 1977).

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| Number | Assumption | Explanation |
| :--- | :--- | :--- |
| 1 | Additive errors | $\boldsymbol{y}=\boldsymbol{y}_{\text {perfect }}+\boldsymbol{\varepsilon}$ |
| 2 | Unbiased model | $\boldsymbol{y}_{\text {perfect }}=y_{\text {mo }}\left(x^{\text {exact }}\right)$ |
| 3 | Zero mean errors | $E[\varepsilon]=0$ |
| 4 | Constant variance | $\operatorname{Var}[\varepsilon]=\sigma_{\varepsilon}^{2}$ |
| 5 | Uncorrelated errors | $\operatorname{cov}\left[\varepsilon_{i} \varepsilon_{j}\right]=0$ for $i \neq j$ |
| 6 | Normal probability distribution |  |
| 7 | Known parameters of the probability density <br> distribution of errors |  |
| 8 | No error in the $S_{i j}$ | $\boldsymbol{S}$ is not a random matrix |
| 9 | No prior information regarding the parameters |  |

Table 1 : Statistical assumptions regarding the measurement errors
The first assumption for the measurement errors is that they are purely additive :

$$
\begin{equation*}
\boldsymbol{y}=\boldsymbol{y}_{\text {perfect }}+\boldsymbol{\varepsilon} \tag{2.1}
\end{equation*}
$$

Here $\boldsymbol{y}_{\text {perfect }}$ represents the vector $(m \times 1)$ of (unknown) errorless measurements ${ }^{2}$. Moreover, measurement errors are assumed to be the realizations of a random variable with a Gaussian distribution with zero mean, that is $E[\varepsilon]=0$ (errors unbiased), $E[$.$] being the$ expected value operator (representing the mean of a large number of realizations of the random variable).

The covariance matrix $(m \times m) \quad \boldsymbol{\psi}=\operatorname{cov}(\boldsymbol{\varepsilon})=E\left[(\boldsymbol{\varepsilon}-E[\boldsymbol{\varepsilon}])(\boldsymbol{\varepsilon}-E[\boldsymbol{\varepsilon}])^{t}\right]=E\left[\varepsilon \varepsilon^{t}\right]$ of error measurements contains on its main diagonal the variance $\sigma_{\varepsilon i}^{2}$ of each measurement that is supposed to stay constant for each time $t_{i}, i=1, \ldots, m$. This variance may be known or not. Finally, measurement errors are assumed uncorrelated (error at time $t_{i}$ independent of error at time $t_{j}\left(E\left[\varepsilon_{i} \varepsilon_{j}\right]=0\right.$ for $\left.i \neq j\right)$ then $\boldsymbol{\psi}$ is a diagonal matrix :

$$
\begin{equation*}
\boldsymbol{\Psi}=\operatorname{cov}(\boldsymbol{\varepsilon})=\operatorname{diag}\left(\sigma_{\varepsilon}^{2}, \cdots, \sigma_{\varepsilon}^{2}, \cdots, \sigma_{\varepsilon}^{2}\right)=\sigma_{\varepsilon}^{2} \boldsymbol{I} \tag{2.2}
\end{equation*}
$$

These data can come from a real experiment or can have been numerically created (in order to test the parameter estimation method), using a mathematical model and adding a

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numerical random noise verifying the preceding assumptions. Now the model and its parameters will be presented.

### 2.2. Build a model of the measured signal, define the parameters and first contact with the sensitivities

The objective of such a model is to give a mathematical expression, noted $y_{m o}(t, \boldsymbol{x})=\eta(t, \boldsymbol{x})$ of the perfect measurements $y_{\text {perfect }}(t)$ mentioned above. This model is a function of the independent variable (time) and of $n$ parameters composing the parameters vector ( $n \times 1$ ) noted $\boldsymbol{x}=\left[x_{1} \ldots x_{n}\right]^{\dagger}$. The model vector $(m \times 1)$ is then given by $\boldsymbol{y}_{m o}(\boldsymbol{t}, \boldsymbol{x})=\left[y_{m o, 1}\left(t_{1}, \boldsymbol{x}\right) \ldots y_{m o, i}\left(t_{i}, \boldsymbol{x}\right) \ldots y_{m o, m}\left(t_{m}, \boldsymbol{x}\right)\right]^{t}$, where $\boldsymbol{t}=\left[t_{1} \ldots t_{i} \ldots t_{m}\right]^{\dagger}$ is a column vector composed of the $m$ times of measurements $t$. For this example, we choose to analyse the classical two parameters estimation problem consisting in estimating simultaneously the slope and the intercept of a straight line; then the model is given, in a scalar writing, by :

$$
\begin{equation*}
y_{m o}(t, \boldsymbol{x})=x_{1} t+x_{2} \tag{2.3}
\end{equation*}
$$

The model is linear with respect to its two parameters $x_{1}$ and $x_{2}$ because

$$
\begin{align*}
& y_{m o}\left(t, a \boldsymbol{x}+b \boldsymbol{x}^{\prime}\right)=\left(a x_{1}+b x_{1}^{\prime}\right) t+a x_{2}+b x_{2}^{\prime}=a\left(x_{1} t+x_{2}\right)+b\left(x_{1}^{\prime} t+x_{2}^{\prime}\right) \\
& y_{m o}\left(t, a \boldsymbol{x}+b \boldsymbol{x}^{\prime}\right)=a \cdot y_{m o}(t, \boldsymbol{x})+b \cdot y_{m o}\left(t, \boldsymbol{x}^{\prime}\right) \tag{2.4}
\end{align*}
$$

Important remark : the following model :

$$
\begin{equation*}
y_{m_{0}}(t, \boldsymbol{x})=x_{1} \sqrt{t}+x_{2} \operatorname{erf}(t) \tag{2.5}
\end{equation*}
$$

is also linear with respect to its two parameters $X_{1}$ and $x_{2}$, even if its time behavior is not linear. On the contrary, the following model :

$$
\begin{equation*}
y_{m o}(t, \boldsymbol{x})=x_{1} \sqrt{t}+\exp \left(-x_{2} t\right) \tag{2.6}
\end{equation*}
$$

is linear with respect to $X_{1}$ but non linear with respect to $x_{2}$ and is consequently non linear with respect to $\boldsymbol{x}$.

Writing the $m$ model values (2.3) for the $m$ time values $t_{1} \ldots t_{m}$, the $m$ resulting equations can be written in a matrix way as follows :

$$
\left[\begin{array}{c}
y_{m o, 1}  \tag{2.7}\\
\vdots \\
y_{m o, i} \\
\vdots \\
y_{m o, m}
\end{array}\right]=\left[\begin{array}{cc}
t_{1} & 1 \\
\vdots & \vdots \\
t_{i} & 1 \\
\vdots & \vdots \\
t_{m} & 1
\end{array}\right]\left[\begin{array}{l}
x_{1} \\
x_{2}
\end{array}\right]
$$

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or, in a more compact form:

$$
\begin{equation*}
\boldsymbol{y}_{m o}=\boldsymbol{S} \boldsymbol{x} \tag{2.8}
\end{equation*}
$$

The matrix $\boldsymbol{S}(m \times n)$ is called the sensitivity matrix. Column $k$ contains the $m$ times values of the sensitivity coefficient of the model with respect to the parameter $x_{k}$, given by :

$$
\begin{equation*}
S_{k}(t, \boldsymbol{x})=\left.\frac{\partial y_{m o}(t, \boldsymbol{x})}{\partial x_{k}}\right|_{t, x_{j} \text { for } j \neq k} \quad, \quad k=1, \ldots, n \tag{2.9}
\end{equation*}
$$

Equation (2. 8) is only valid for a linear model. However, the sensivity coefficient (2.9a) can be defined for the discrete time values $t=t_{i}(i=1, \ldots, m)$ to form a sensitivity matrix $\boldsymbol{S}$ defined for any linear or non linear model as:

$$
\begin{equation*}
\boldsymbol{S}(\boldsymbol{x})=\left[\nabla_{x} \boldsymbol{y}_{m o}(\boldsymbol{x})\right]^{t} \text { or, more simply, in a symbolic way } \boldsymbol{S}(\boldsymbol{x})=\frac{\mathrm{d} \boldsymbol{y}_{m o}(\boldsymbol{x})}{\mathrm{d} \boldsymbol{x}} \tag{2.10}
\end{equation*}
$$

Important remark : if the model is linear with respect to its parameters (as in the cases (2. 3) and (2.5)), then the sensitivity coefficients do not depend on parameters: $S_{k}(t, \boldsymbol{x})=S_{k}(t)$, and the sensitivity matrix $\boldsymbol{S}(\boldsymbol{x})$ does not depend on $\boldsymbol{x}$.

For the model (2. 3), we have $S_{1}(t)=t$ and $S_{2}(t)=1$ then

$$
\boldsymbol{S}=\left[\begin{array}{cc}
S_{1}\left(t_{1}\right) & S_{2}\left(t_{1}\right)  \tag{2.11}\\
\vdots & \vdots \\
S_{1}\left(t_{i}\right) & S_{2}\left(t_{i}\right) \\
\vdots & \vdots \\
S_{1}\left(t_{m}\right) & S_{2}\left(t_{m}\right)
\end{array}\right]=\left[\begin{array}{cc}
t_{1} & 1 \\
\vdots & \vdots \\
t_{i} & 1 \\
\vdots & \vdots \\
t_{m} & 1
\end{array}\right]
$$

A sensitivity coefficient is a measure of the "influence" of parameter $x_{k}$ on the response of the model $\boldsymbol{y}_{\boldsymbol{m o}}(\boldsymbol{t}, \boldsymbol{x})$. If all the sensitivity coefficients are of "high" magnitude and "independent", the simultaneous estimation of the parameters composing $\boldsymbol{x}$ will be possible. The meaning of "high" and "independent" will be developed later.

### 2.3. Choose the objective function

Assuming that the model has the right form (or "right structure", given par the resolution of the "right" partial differential equations describing the "right" physical phenomena) and is calculated with the right values of parameters $\boldsymbol{x}^{\text {exact }}$, then $\boldsymbol{y}_{\text {mo }}\left(\boldsymbol{x}^{\text {exact }}\right)=\boldsymbol{y}_{\text {perfect }}$ and (2. 1) becomes

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$$
\begin{equation*}
\boldsymbol{y}=\boldsymbol{y}_{m o}\left(\boldsymbol{x}^{\text {exact }}\right)+\boldsymbol{\varepsilon} \tag{2.12}
\end{equation*}
$$

Since the $m$ measurement errors composing $\varepsilon$ are not known, the problem of finding the values of the $n$ components of $\boldsymbol{x}^{\text {exact }}$ given $m$ measurements verifying (2. 12) is underdetermined ( $m$ equations with $n+m$ unknowns: $n$ parameters $x_{k} \quad(k=1, \ldots, n$ ) and $m$ noise values $\left.\varepsilon_{i}(i=1, \ldots, m)\right)$. The problem consists in using the $m$ measurements for estimating the $n$ unknown parameters, with $m \geq n$. Then the new problem to solve is a minimization problem. For a given value $\boldsymbol{x}$ of the parameter vector, a residual vector $\boldsymbol{r}(m \times 1)$ is built in order to calculate the difference between measurement vector $\boldsymbol{y}(m \times 1)$ and the corresponding model output $\boldsymbol{y}_{m o}(\boldsymbol{x})(m \times 1)$, each component of $\boldsymbol{r}$ being associated with one of the $m$ instants of time where a measurement is available.

$$
\boldsymbol{r}(\boldsymbol{x})=\boldsymbol{y}-\boldsymbol{y}_{m o}(\boldsymbol{x})=\left[\begin{array}{lllll}
y_{1}-y_{m o, i}\left(t_{1}, \boldsymbol{x}\right) & \ldots & y_{i}-y_{m o, i}\left(t_{i}, \boldsymbol{x}\right) & \ldots & y_{m}-y_{m o, m}\left(t_{m}, \boldsymbol{x}\right) \tag{2.13}
\end{array}\right]^{t}
$$

This present definition of the residual vector $\boldsymbol{r}(\boldsymbol{x})$ is an extension of the concept of residual vector which is usually defined as $\boldsymbol{r}(\hat{\boldsymbol{x}})$, where $\hat{\boldsymbol{x}}$ corresponds to the minimum of the square of the norm of $\boldsymbol{r}(\boldsymbol{x})$, see equation (2.14) further on.

Then the norm of this residual vector $\|\boldsymbol{r}(\boldsymbol{x})\|$ is calculated, it is a scalar value that will be minimized with respect to the different components of parameter $\boldsymbol{x}$ in order to estimate an 'opimal' value for it. One have to choose the way of computing the norm of the residuals $\|\boldsymbol{r}(\boldsymbol{x})\|$. Without any a priori information on the parameters and given the above assumptions for measurements errors, the chosen norm is the Euclidian norm (or $\mathrm{L}_{2}$ norm) given by

$$
\begin{equation*}
\|\boldsymbol{r}(\boldsymbol{x})\|=\left(\sum_{i=1}^{m} r_{i}(\boldsymbol{x})^{2}\right)^{1 / 2} \tag{2.14}
\end{equation*}
$$

In fact the objective function that will be minimized is the square of thist Euclidian norm, it is called the 'Ordinary Least Squares' objective function ${ }^{3}$ :

$$
\begin{equation*}
J_{O L S}(\boldsymbol{x})=\|\boldsymbol{r}(\boldsymbol{x})\|^{2}=\left\|\boldsymbol{y}-\boldsymbol{y}_{m o}(\boldsymbol{x})\right\|^{2} \tag{2.14}
\end{equation*}
$$

In the particular case of a linear model, $\boldsymbol{y}_{m o}(\boldsymbol{x})=\boldsymbol{S} \boldsymbol{x}$ and this OLS sum becomes :

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$$
\begin{equation*}
\left.J_{O\llcorner S}(\boldsymbol{x})=\sum_{i=1}^{m} r_{i}(\boldsymbol{x})^{2}=\sum_{i=1}^{m}\left(y_{i}-y_{m o, i}\left(t_{i}, \boldsymbol{x}\right)\right)^{2}=\sum_{i=1}^{m}\left(y_{i}-\sum_{j=1}^{n} S_{j}\left(t_{i}\right) \hat{x}_{j}\right)\right)^{2} \tag{2.15}
\end{equation*}
$$

With a matrix writing, (2.15) is equivalent to :

$$
\begin{equation*}
J_{O L S}(\boldsymbol{x})=\left[\boldsymbol{y}-\boldsymbol{y}_{m o}(\boldsymbol{x})\right]^{t}\left[\boldsymbol{y}-\boldsymbol{y}_{m o}(\boldsymbol{x})\right] \tag{2.16}
\end{equation*}
$$

The solution of this minimization will be called $\hat{\boldsymbol{x}}_{\text {OLS }}$ here. The hat ( $\wedge$ ) superscript designates an estimator of the corresponding quantity, that is a random variable deriving here from the random vector variable $\boldsymbol{\varepsilon}$ (the measurement noise) and the subscript 'OLS' designates the specific minimized norm used here, the Ordinary Least Squares sum $J_{\text {oLS }}$ defined in equation (2.14). If the model is linear, this OLS estimator does not require the use of any iterative algorithm and is given in a simple explicit form:

$$
\begin{equation*}
\hat{\boldsymbol{x}}_{O L S}=\arg \left[\min \left(J_{O L S}(\boldsymbol{x})\right)\right] \tag{2.17}
\end{equation*}
$$

So, the original questions was:
"what are the exact values $\boldsymbol{x}^{\text {exact }}$ of parameter vector $\boldsymbol{x}$ for the model $\boldsymbol{y}_{m o}(\boldsymbol{x})$ when $m$ corresponding noisy measurements $\boldsymbol{y}=\boldsymbol{y}_{m o}\left(\boldsymbol{x}^{\text {exact }}\right)+\boldsymbol{\varepsilon}$ are available ?"

The answer is :
"one possible approximation of $\boldsymbol{x}^{\text {exact }}$ is the estimator $\hat{\boldsymbol{x}}_{\text {OLS }}$, which minimizes the Ordinary Least Squares 'objective' function (sometimes also called 'criterion') $J_{O L S}(\boldsymbol{x})$ defined as the sum of the squares of the differences between the $m$ model output and the corresponding measurements".

Or, in simpler words:
"the natural numerical approximation of the parameters present in $\boldsymbol{x}^{\text {exact }}$ is the one that enables the model to be the closest to the whole set of measurements". This Ordinary Least Squares method was first found by Carl Friedrich Gauss in 1795 and later published by Adrien-Marie Legendre in 1805.

The natural question that arises next is: "how far is this $\hat{\boldsymbol{x}}_{\text {OLS }}$ estimation from the exact value $\boldsymbol{x}^{\text {exact }}$ and what can be done to reduce their difference ?" These questions will be discussed now within the linear assumption where an explicit expression for $\hat{\boldsymbol{x}}_{\text {OLS }}$ will be given. Readers interested by non linear estimation can refer to lecture 4 of this series.

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### 2.4. Solve the parameter estimation problem : minimize the objective function

The OLS estimator $\hat{\boldsymbol{x}}_{\text {OLS }}$ is defined as the value of parameter vector $\boldsymbol{x}$ which minimizes the scalar function $J_{O L S}(\boldsymbol{x})$. So, it has to verify :

$$
\begin{align*}
& \nabla_{x} J_{\text {OLS }}\left(\hat{\boldsymbol{x}}_{\text {OLS }}\right)=0 \text { with } \nabla_{x}=\left(\begin{array}{c}
\frac{\partial}{\partial x_{1}} \\
\vdots \\
\frac{\partial}{\partial x_{n}}
\end{array}\right)  \tag{2.18}\\
& \nabla_{x} J_{\text {OLS }}(\boldsymbol{x})=2\left[\nabla_{x}\left[\boldsymbol{y}-\boldsymbol{y}_{m o}(\boldsymbol{x})\right]\right]^{t}\left[\boldsymbol{y}-\boldsymbol{y}_{m o}(\boldsymbol{x})\right] \tag{2.19}
\end{align*}
$$

Knowing that

$$
\begin{equation*}
\boldsymbol{S}^{t}=\left[\nabla_{x} \boldsymbol{y}_{m o}(\boldsymbol{x})\right]^{t} \text { and } \boldsymbol{y}_{m o}(\boldsymbol{x})=\boldsymbol{S} \boldsymbol{x} \tag{2.20}
\end{equation*}
$$

(2.19) becomes

$$
\begin{equation*}
\nabla_{\chi} J_{O L S}(\boldsymbol{x})=-2 \boldsymbol{S}^{t}[\boldsymbol{y}-\boldsymbol{S} \boldsymbol{x}] \tag{2.21}
\end{equation*}
$$

Then $\hat{\boldsymbol{x}}_{\text {oLS }}$ is solution of

$$
\begin{equation*}
\left[\boldsymbol{S}^{t} \boldsymbol{S}\right] \boldsymbol{x}=\boldsymbol{S}^{t} \boldsymbol{y} \tag{2.22}
\end{equation*}
$$

The $n$ equations composing the linear system (2.22) are called the 'normal equations'. The solution is straightforward if the $(n \times n)$ matrix $\boldsymbol{S}^{t} \boldsymbol{S}$ is not singular, it is then possible to compute its inverse and to obtain :

$$
\begin{equation*}
\hat{\boldsymbol{x}}_{O L S}=\left[\boldsymbol{S}^{t} \boldsymbol{S}\right]^{-1} \boldsymbol{S}^{t} \boldsymbol{y} \tag{2.23}
\end{equation*}
$$

The ( $n \times m$ ) matrix $\left[\boldsymbol{S}^{t} \boldsymbol{S}\right]^{-1} \boldsymbol{S}^{t}$ is called the Moore-Penrose matrix, or the pseudo-inverse of $\boldsymbol{S}$. Obviously, a necessary condition for $\boldsymbol{S}^{t} \boldsymbol{S}$ to be non singular is that the sensitivity coefficients are independent, and have a non zero norm. This condition also requires that the number of measurements $m$ be equal or greater than the number of parameters $n$ to be estimated.

Equation (2.22) gives an explicit expression for the ordinary least square estimator $\hat{\boldsymbol{x}}_{\mathcal{O L S}}$ of $\boldsymbol{x}$ for any linear model $\boldsymbol{y}_{m o}(\boldsymbol{x})=\boldsymbol{S} \boldsymbol{x}$ as a function of measurements $\boldsymbol{y}$ defined in equation (2.11). Since $\boldsymbol{y}$ is a random vector (because of noise $\boldsymbol{\varepsilon}$ ), such is also the case for $\hat{\boldsymbol{x}}_{\text {OLS }}$. However, equation (2.22) has also another statistical meaning: once measurements are available, a realization of $\boldsymbol{y}$ (that is numerical values for its components) becomes available, and this equation provides the corresponding OLS estimation of $\boldsymbol{x}$.

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### 2.5. Evaluate the confidence in estimations (variance and bias of estimator)

### 2.5.1. First approach with stochastic simulations (Monte Carlo method)

Before computing the statistical properties of the OLS estimator (expected value and covariance matrix), we present a graphical approach that helps to understand the meaning of such properties. This approach is possible in the case when two parameters are estimated because each estimation $\hat{\boldsymbol{x}}_{\text {OLS }}=\left(\hat{x}_{\text {OLS, },}, \hat{x}_{\text {OLS, } 2}\right)$ can be plotted as a point in a 2 D coordinates axes graduated in $\left(x_{1}, x_{2}\right)$. The idea is then to simulate $K=100$ experiments with $K$ different realizations of the random noise vector $\boldsymbol{\varepsilon}$ with the same statistical properties (see Table 1) to produce $K$ samples of measurements vectors $\boldsymbol{y}$ according to (2. 12). The conditions of each simulated experiment that is called 'the reference case' are indicated in Table 2. Figure 1 shows one of the simulated experiment (circles) and one of the corresponding estimated model calculated with the OLS estimation $\hat{\boldsymbol{x}}_{\text {OLS }}=\left(\hat{x}_{O L S, 1}, \hat{x}_{O L S, 2}\right)$ (red line).

| $x_{1}^{\text {exact }}$ | 5 |
| :--- | :--- |
| $x_{2}^{\text {exact }}$ | 2 |
| Model structure $y_{m o}(t, \boldsymbol{x})$ | $x_{1} t+x_{2}$, eq (2.3) |
| Number of measurements $m$ | 20 |
| Start of time range $\mathrm{t}_{\text {min }}$ | 0.5 |
| Time step dt | 0.1 |
| Noise standard deviation $\sigma_{\varepsilon}$ | 0.5 |

Table 2 : conditions of the $K=100$ simulated 'reference' experiments

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Figure 1 : one of the $K=100$ experiments of the 'reference case', the corresponding exact model and the corresponding OLS model

The $K=100$ OLS estimations $\hat{\boldsymbol{x}}_{\text {OLS }}=\left(\hat{X}_{\text {OLS, } 1}, \hat{x}_{\text {OLS, } 2}\right)$ are then plotted in a scatter graph graduated in ( $x_{1}, x_{2}$ ) in Figure 2. Because of a different random realization of noise for each of the 100 experiments, each corresponding OLS estimations $\hat{\boldsymbol{x}}_{\text {OLS }}=\left(\hat{x}_{O L S, 1}, \hat{x}_{O L S, 2}\right)$ is different, showing immediately the consequence of noise measurement on the dispersion of estimations.

In that figure, the position (square) of the exact value $\boldsymbol{x}^{\text {exact }}=\left(x_{1}^{\text {exact }}=5, x_{2}^{\text {exact }}=2\right)$ and the position (star) of the mean value of the $K$ estimations $\hat{\boldsymbol{x}}_{\text {mean }}=\left(\right.$ mean $\left.\left(\hat{X}_{1, \text { OLS }}\right)=4.994, \operatorname{mean}\left(\hat{X}_{2, \text { OLS }}\right)=2.019\right)$ (the center of the scatter, are very close to their exact values).

Another interesting way of looking at the estimation results is to plot them in a scatter graph with normalized coordinates indicating the distance of each estimation from the center of the scatter in \%, see Figure 3 :

$$
\begin{align*}
& e_{O L S, 1, i}=100\left(\hat{x}_{\text {OLS, } 1, i}-\hat{x}_{\text {mean }, 1}\right) / \hat{x}_{\text {mean }, 1}  \tag{2.24}\\
& e_{O L S, 2, i}=100\left(\hat{x}_{O L S, 2, i}-\hat{x}_{\text {mean }, 2}\right) / \hat{x}_{\text {mean }, 2} \tag{2.25}
\end{align*}
$$

If we consider that $\hat{\boldsymbol{x}}_{\text {mean }} \approx \boldsymbol{X}_{\text {exact }}$ the quantities (2.24) and (2.25) are the relative estimation errors for $x_{1}^{\text {exact }}$ and $x_{2}^{\text {exact }}$. This plot enables to quantify, in $\%$, the dispersion of the estimations of each parameter around its mean value. That is often what is wanted to be minimized.

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Figure 2 : dispersion of the 100 estimations around their central value (star) that is very near the exact value (square)


Figure 3 : errors of estimations in \% of the mean value of the scatter

At this point, after having quantified the central value $\hat{\boldsymbol{x}}_{\text {mean }}$ of the $K=100$ $\hat{\boldsymbol{x}}_{\text {OLS }}$ estimations and after having evaluated the dispersion of the majority of estimations around their central values (which indicate the confidence we associate to them), we can sum up the result of the estimation problem in the following way :

$$
\begin{aligned}
" x_{1}^{\text {exact }} \text { is equal to } \hat{x}_{\text {mean } 1} & =2.0 \pm 20 \% \text { and } \\
x_{2}^{\text {exact }} \text { is equal to } \hat{x}_{\text {mean } 2} & =4.9 \pm 40 \% "
\end{aligned}
$$

But in reality, we never realize 100 experiments with 100 estimations $\hat{\boldsymbol{x}}_{\text {OLS }}=\left(\hat{x}_{\text {OLS, } 1}, \hat{x}_{\text {OLS, } 2}\right)$ in order to calculate the mean value $\hat{\boldsymbol{x}}_{\text {mean }}=\left(\operatorname{mean}\left(\hat{x}_{1}\right), \operatorname{mean}\left(\hat{x}_{2}\right)\right)$. We generally do one single experiment, and obtain only one of the 100 points of Figure 2 and Figure 3.

We must keep in mind that this point can be one of the points 'far' from the exact value !
Whatever the realized experiment among these 100, what we want to do is to associate a 'confidence region' to the particular estimation $\hat{\boldsymbol{x}}_{\text {OLS }}=\left(\hat{x}_{O L S, 1}, \hat{x}_{O L S, 2}\right)$ (or a 'confidence interval' for each parameter) that has about the same dimension than the scatter we have just obtained with these 100 simulated experiments. That is the objective of the following section.

### 2.5.2. Calculation of statistical properties of the OLS estimator

Here we become more general and we consider the case when not all the $n$ parameters are estimated but only $r$, the ( $n-r$ ) remaining parameters are supposed to be known and they are fixed during the estimation of the $r$ unknown parameters. Estimated parameters are noted with subscript $r$ and fixed parameters are noted with subscript $c$. We must consider that the fixed parameters have not been fixed to their exact value, and at the end of estimation of the

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$r$ parameters, we have to evaluate the bias made on estimations because of the error on the $(n-r)$ parameters that are supposed to be known ( $n-r$ ).

We can split (2.8) into

$$
\begin{equation*}
\boldsymbol{y}_{m o}=\boldsymbol{S}_{r} \boldsymbol{x}_{r}+\boldsymbol{S}_{c} \boldsymbol{x}_{c} \tag{2.26}
\end{equation*}
$$

Matrix $\boldsymbol{S}_{r}(n \times r)$ is the sensitivity matrix to the parameters to be estimated. It is part of the "complete" sensitivity matrix $\boldsymbol{S}$, relative to all the parameters (to be estimated $\boldsymbol{x}_{r}(r \times 1)$ and fixed $\left.\boldsymbol{x}_{c}((n-r) \times 1)\right)$ :

$$
\left.\boldsymbol{S}=\left[\boldsymbol{S}_{r} \vdots \boldsymbol{S}_{c}\right]=\left[\begin{array}{ccc}
S_{1}\left(t_{1}\right) & \ldots & S_{r}\left(t_{1}\right)  \tag{2.27}\\
\vdots & \ldots & \vdots \\
S_{1}\left(t_{m}\right) & \ldots & S_{r}\left(t_{m}\right)
\end{array}\right] \vdots \vdots\left[\begin{array}{ccc}
S_{r+1}\left(t_{1}\right) & \ldots & S_{n-r}\left(t_{1}\right) \\
\vdots & \ldots & \vdots \\
S_{r+1}\left(t_{m}\right) & \ldots & S_{n-r}\left(t_{m}\right)
\end{array}\right]\right]
$$

Matrix $\boldsymbol{S}_{c}(\mathrm{n} \times r)$ is the sensitivity matrix to estimated parameters. It is a part of the "complete" sensitivity matrix $\boldsymbol{S}$, relative to all the parameters (estimated $\boldsymbol{x}_{r}(r \times 1)$ and fixed $\left.\boldsymbol{x}_{c}((n-r) \times 1)\right)$ :

The OLS solution (2.23) becomes

$$
\begin{equation*}
\hat{\boldsymbol{x}}_{O L S}=\left[\boldsymbol{S}_{r}^{t} \boldsymbol{S}_{r}\right]^{-1} \boldsymbol{S}_{r}^{t}\left(\boldsymbol{y}-\boldsymbol{S}_{c} \boldsymbol{x}_{c}\right) \tag{2.28}
\end{equation*}
$$

Let $\hat{\boldsymbol{x}}_{r, \text { ols }}\left(\tilde{\boldsymbol{x}}_{c}\right)$ be the estimated parameters for a value of fixed parameters $\tilde{\boldsymbol{x}}_{c}$ different from their exact value $\boldsymbol{x}_{c}^{\text {exact }}$. Let $\boldsymbol{e}_{r}$ be the vector ( $r \times 1$ ) of error estimation (the difference between estimated and exact values of $\boldsymbol{x}_{r}^{\text {exact }}$ ) and let $\boldsymbol{e}_{c}$ be the deterministic error (the bias) for the values of the fixed parameters:

$$
\begin{align*}
& \boldsymbol{e}_{r}=\hat{\boldsymbol{x}}_{r, \text { oLs }}\left(\tilde{\boldsymbol{x}}_{c}\right)-\boldsymbol{X}_{r}^{\text {exact }}  \tag{2.29}\\
& \boldsymbol{e}_{c}=\tilde{\boldsymbol{x}}_{c}-\boldsymbol{x}_{c}^{\text {exact }} \tag{2.30}
\end{align*}
$$

One can write, with $\boldsymbol{A}_{r}=\left[\mathbf{S}_{r}^{t} \mathbf{S}_{r}\right]^{-1} \mathbf{S}_{r}^{t}$ the Moore-Penrose matrix

$$
\begin{equation*}
\hat{\boldsymbol{x}}_{r, \text { OLS }}\left(\tilde{\boldsymbol{x}}_{c}\right)=\boldsymbol{A}_{r}\left(\boldsymbol{y}-\boldsymbol{S}_{c} \tilde{\boldsymbol{x}}_{c}\right) \tag{2.31}
\end{equation*}
$$

Equation (2. 12) can be developed:

$$
\begin{equation*}
\boldsymbol{y}=\boldsymbol{y}_{m o}\left(\boldsymbol{x}^{\text {exact }}\right)+\boldsymbol{\varepsilon}=\boldsymbol{S}_{r} \boldsymbol{x}_{r}^{\text {exact }}+\boldsymbol{S}_{c} \boldsymbol{x}_{c}^{\text {exact }}+\boldsymbol{\varepsilon} \tag{2.32}
\end{equation*}
$$

Combining (2.32) and (2.31), the estimation error of (2.29) may then be approached by :

$$
\begin{equation*}
\boldsymbol{e}_{r}=\hat{\boldsymbol{x}}_{r, \text { OLS }}\left(\tilde{\boldsymbol{x}}_{c}\right)-\boldsymbol{x}_{r}^{\text {exact }}=\boldsymbol{A} \boldsymbol{\varepsilon}-\boldsymbol{A}_{r} \boldsymbol{S}_{c} \boldsymbol{e}_{c}=\boldsymbol{e}_{r 1}+\boldsymbol{e}_{r 2} \tag{2.33}
\end{equation*}
$$

The first term $\boldsymbol{e}_{r 1}=\boldsymbol{A}_{t} \boldsymbol{\varepsilon}$ is the random contribution to the total error; it represents the error due to measurement errors $\boldsymbol{\varepsilon}$ whose covariance matrix $\boldsymbol{\Psi}$ is given by (2. 2). The second term $\boldsymbol{e}_{r 2}=-\boldsymbol{A}_{r} \boldsymbol{S}_{c} \boldsymbol{e}_{c}$ is the non-random (deterministic) contribution to the total error vector due to the deterministic error on the fixed parameters $\boldsymbol{e}_{c}$. The expected value of $\boldsymbol{e}_{r 1}$ is

$$
\begin{equation*}
E\left[\boldsymbol{e}_{r 1}\right]=\boldsymbol{A}_{t} E[\boldsymbol{\varepsilon}]=0 \tag{2.34}
\end{equation*}
$$

meaning that no systematic bias is introduced by the random measurement errors.
Remark : this explains that the mean $\hat{\boldsymbol{x}}_{\text {mean }}$ of the 100 scattered estimations in Figure $\mathbf{2}$ is very close to the exact value $\boldsymbol{x}^{\text {exact }}$.

The covariance matrix of $\boldsymbol{e}_{r 1}$ is given by

$$
\begin{equation*}
\boldsymbol{C}_{1}=\operatorname{cov}\left(\boldsymbol{e}_{r 1}\right)=E\left[\boldsymbol{e}_{r 1} \boldsymbol{e}_{r 1}^{t}\right]=\boldsymbol{A}_{r} E\left[\boldsymbol{\varepsilon} \boldsymbol{\varepsilon}^{t}\right] \boldsymbol{A}_{r}^{t}=\boldsymbol{A}_{r} \boldsymbol{\Psi} \boldsymbol{A}_{r}^{t}=\left[\boldsymbol{S}_{r}^{t} \mathbf{S}_{r}\right]^{-1} \sigma_{\varepsilon}^{2} \tag{2.35}
\end{equation*}
$$

Matrix $\left[\mathbf{S}_{r}^{t} \mathbf{S}_{r}\right]^{-1}$ may thus be seen as a matrix that "amplifies" the measurement errors. The expected value of $\boldsymbol{e}_{r 2}$ is

$$
\begin{equation*}
E\left[\boldsymbol{e}_{r 2}\right]=-\boldsymbol{A}_{r} \boldsymbol{S}_{c} \boldsymbol{e}_{c}=\left[\boldsymbol{S}_{r}^{t} \boldsymbol{S}_{r}\right]^{-1} \boldsymbol{S}_{r}^{t} \mathbf{S}_{c} \boldsymbol{e}_{c} \neq 0 \tag{2.36}
\end{equation*}
$$

This expected value is different from zero, which means that estimation $\hat{\boldsymbol{x}}_{r, \text { ols }}$ is biased, if the error $\boldsymbol{e}_{c}$ of the parameters supposed to be known is different from zero itself. This means that in the preceding stochastic simulation, the scatter of 100 estimations would not have been centred on $\boldsymbol{x}_{r}^{\text {exact }}$. This bias is computed using the corresponding sensitivity coefficients matrix $\boldsymbol{S}_{c}$. The covariance matrix $((n-r) \times(n-r))$ of $\boldsymbol{e}_{r 2}$ error is $\boldsymbol{C}_{2}=\operatorname{cov}\left(\boldsymbol{e}_{r 2}\right)=0$ because $\boldsymbol{e}_{c}$ is not a random error. Finally, the total bias associated to the estimation $\hat{\boldsymbol{x}}_{r, O L S}\left(\tilde{\boldsymbol{x}}_{c}\right)$ is due to the biased value of $\tilde{\boldsymbol{x}}_{c}$ and its value is given by

$$
\begin{equation*}
E\left[\boldsymbol{e}_{r}\right]=E\left[\boldsymbol{e}_{r 2}\right]=-\boldsymbol{A}_{r} \mathbf{S}_{c} \boldsymbol{e}_{c}=\left[\boldsymbol{S}_{r}^{t} \mathbf{S}_{r}\right]^{-1} \boldsymbol{S}_{r}^{t} \mathbf{S}_{c} \boldsymbol{e}_{c} \tag{2.37}
\end{equation*}
$$

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The matrix $\left[\boldsymbol{S}_{r}^{t} \mathbf{S}_{r}\right]^{-1} \mathbf{S}_{r}^{t} \boldsymbol{S}_{c}(r \times(n-r))$ may thus be seen as the "amplification" of bias on fixed parameters $\boldsymbol{e}_{c}$. For a fixed value of $\tilde{\boldsymbol{x}}_{c}$, the covariance matrix $\boldsymbol{C}_{r}$ of estimations errors is

$$
\begin{equation*}
\boldsymbol{C}_{r}=\operatorname{cov}\left(\boldsymbol{e}_{r}\right)=E\left[\left(\boldsymbol{e}_{r}-E\left[\boldsymbol{e}_{r}\right]\right)\left(\boldsymbol{e}_{r}-E\left[\boldsymbol{e}_{r}\right]\right)^{t}\right]=E\left[\boldsymbol{e}_{r 1} \boldsymbol{e}_{r 1}^{t}\right]=\operatorname{cov}\left(\boldsymbol{e}_{r 1}\right)=\boldsymbol{C}_{1} \tag{2.38}
\end{equation*}
$$

The coefficients of the covariance matrix are :

$$
\boldsymbol{C}_{r}=\left[\begin{array}{cccc}
\sigma_{1}^{2} & \operatorname{cov}\left(e_{r 1}, e_{r 2}\right) & \ldots & \operatorname{cov}\left(e_{r 1}, e_{r r}\right)  \tag{2.39}\\
& \sigma_{2}^{2} & & \operatorname{cov}\left(e_{r 2}, e_{r r}\right) \\
& & \ddots & \vdots \\
\operatorname{sym} & & & \sigma_{r}^{2}
\end{array}\right]
$$

Its main diagonal elements are the individual variances of the errors associated to each component of the estimated vector $\hat{\boldsymbol{x}}_{r, \text { OLS }}$. Its off diagonal coefficients are the covariances of the crossed errors. Expression (2.35) shows that knowledge of the variance of the measurement errors $\sigma_{\varepsilon}^{2}$ is needed in order to compute the covariance matrix of the estimation errors. If $\sigma_{\varepsilon}^{2}$ is not measured before the experiment, an estimation of it may be obtained at the end of estimation thanks to the minimum value of the objective function: $J_{O L S}\left(\hat{\boldsymbol{x}}_{r, O L S}\left(\tilde{\boldsymbol{x}}_{c}\right)\right)=\sum_{i=1}^{m} r_{i}\left(\hat{\boldsymbol{x}}_{O L S}\left(\tilde{\boldsymbol{x}}_{c}\right)\right)^{2}$.

In fact, this estimation is based on the fact that, at the end of the estimation, the only difference that subsists between measurements and model (if its structure and its parameters are correct) must be the measurement errors. Of course, exact parameters can not be exactly obtained, and the remaining differences between measurements and model are the cause of non zero residuals given by (2. 13). If estimated parameters are not too far from exact parameters, residuals must have some statistical properties close to those of measurement errors. That is why a non biased estimation of $\sigma_{\varepsilon}^{2}$ for the estimation of $r$ parameters from the use of $m$ measurements is thus given by

$$
\begin{equation*}
\hat{\sigma}_{\varepsilon}^{2}=\frac{J_{\text {OLS }}\left(\hat{\boldsymbol{x}}_{r, \text { OLS }}\left(\tilde{\boldsymbol{x}}_{c}\right)\right)}{n-r} \tag{2.40}
\end{equation*}
$$

This estimation is only valid for errorless values of the parameters that are supposed to be known, that is for $\tilde{\boldsymbol{x}}_{c}=\boldsymbol{x}_{c}^{\text {exact }}$, for an i.i.d. (independent and identically distributed) noise $\boldsymbol{\varepsilon}$ which corresponds to assumptions 1 to 5 and 8 to 9 in Table 1.

### 2.5.3. The correlation matrix

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The variance $\sigma_{i}^{2}$ of the estimation error associated to $\hat{\boldsymbol{x}}_{r, \text { OLS }}(i)$ can not be arbitrarily low independently of $\sigma_{j}^{2}(j \neq i)$ if $\operatorname{cov}\left(e_{r i}, e_{r j}\right) \neq 0$ because $\sigma_{i} \sigma_{j} \geq \operatorname{cov}\left(e_{r i}, e_{r j}\right)$ (Schwarz relationship) : $\hat{x}_{r, i}$ and $\hat{x}_{r, j}$ are said correlated. The correlation level between estimations $\hat{x}_{r, i}$ and $\hat{x}_{r, j}$ is thus measured by the quantity

$$
\begin{equation*}
\rho_{i j}=\frac{\operatorname{cov}\left(e_{r i}, e_{i j}\right)}{\sigma_{i} \sigma_{j}}=\frac{C_{r, i j}}{\sqrt{C_{r, i} C_{r, j i}}}=\frac{P_{r, i j}}{\sqrt{P_{r, i j} P_{r, j i}}}, i, j=1, \ldots, r \tag{2.41}
\end{equation*}
$$

that lies between -1 and 1 . These coefficients compose the $\boldsymbol{V}_{C O R}$ matrix ( $r \times r$ ). One considers that the estimations are highly correlated when $\left|\rho_{i j}\right| \geq 0.9$ (Beck et al., 1977). This quantity is independent of the magnitude of measurement errors. In the example of Figure 3, $\rho_{12}=0.99$ indicates that the accuracy for the estimation of the slope $\left(X_{1}\right)$ is highly linked to the accuracy for the estimation of the intercept. If the variance on the slope is low, then the variance on the intercept will be high. This is why the scatter of the 100 estimations is contained inside a 'narrow' and 'inclined' ellipse. The ideal conditions would be that the accuracy for the estimation of all parameters be independent of each other. That gives an idea for a criterion to respect when designing an experiment for estimating several parameters.

### 2.5.4. The confidence region and interval for OLS with standard assumptions

In that case, with the additional assumption 6 (see Table 1) of a Gaussian noise, one can show that the confidence region in the plane ( $\hat{x}_{1}, \hat{x}_{2}$ ) of the Figure 2 for a given confidence level of $\alpha$ is an ellipse (for $n=2$ parameters, see Figure 4). Its equation in $\boldsymbol{\delta} \boldsymbol{x}$ coordinates centered on $\hat{\boldsymbol{x}}_{\text {OLS }}$ is:

$$
\begin{align*}
& \boldsymbol{\delta} \boldsymbol{x}^{t} \cdot \boldsymbol{S}^{\mathbf{t}} \boldsymbol{S} \boldsymbol{\delta} \boldsymbol{x}=\Delta^{2} \\
& \Delta^{2}=\chi_{1-\alpha}^{2}(2) \sigma_{\varepsilon}^{2} \tag{2.42}
\end{align*}
$$

$\chi_{1-\alpha}^{2}(2)$ is computed by the function chi2inv(alpha,2) in MATLAB® if we search for the confidence region at a level $95 \%(\alpha=0.95) \cdot \sigma_{\varepsilon}^{2}$ is the variance of noise measurements. It is worth noting that the length of half axes $\rho_{1}$ and $\rho_{2}$ in the principal directions of the ellipse are given by

$$
\begin{align*}
& \rho_{1}=\Delta / \sqrt{\lambda_{1}} \\
& \rho_{2}=\Delta / \sqrt{\lambda_{2}} \tag{2.43}
\end{align*}
$$

$\lambda_{1}$ and $\lambda_{2}$ are the eigenvalues of $\boldsymbol{S}^{t} \boldsymbol{S}$. The product of that two eigenvalues is equal to the determinant of $\boldsymbol{S}^{t} \boldsymbol{S}$. Finally, it is shown that the area of the confidence region inside the ellipse is given by

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$$
\begin{equation*}
A=\pi . \rho_{1} . \rho_{2}=\frac{\pi \chi_{1-\alpha}^{2}(2) \sigma_{\varepsilon}^{2}}{\sqrt{\operatorname{det}\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)}}=\frac{\pi \chi_{1-\alpha}^{2}(2) \sigma_{\varepsilon}^{2}}{\sqrt{\lambda_{1} \lambda_{2}}} \tag{2.44}
\end{equation*}
$$



Figure 4 : elliptical confidence region associated to the estimation of two parameters (with standard assumptions on measurement noise), at a confidence level $\alpha=0.95$.

Then, the product of eigenvalues of $\boldsymbol{s}^{\boldsymbol{t} \boldsymbol{S}}$ gives information on the area of the confidence region, while the individual eigenvalues give information on the lengths of each principal direction of the ellipse: a 'long' ellipse in a direction corresponds to a low eigenvalue. The experiment that will maximize $\operatorname{det}\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)=\lambda_{1} \lambda_{2}$ in order to minimize the confidence region is called a 'D-optimal' experiment.

In the case of estimation of $r=n$ parameters, the matrix $\boldsymbol{C}$ (eq 2.39) contains on its main diagonal the $n$ variances associated to each component of the estimated vector $\hat{\boldsymbol{x}}_{\text {OLS }}$. The square root of the $\mathrm{i}^{\text {th }}$ diagonal component of $\boldsymbol{C}$ is then the standard deviation associated to the estimation $\hat{x}_{i, O L S}$ and can be expressed in \%. Then, the half width of confidence interval $\mathrm{Cl}_{i}^{1-\alpha}$, at a level of confidence of $100(1-\alpha) \%$, associated to the estimation $\hat{x}_{i, \text { OLS }}$ is now given by :

$$
\begin{equation*}
C l_{i}^{1-\alpha}=t_{1-\alpha / 2}(n-r) \times \sqrt{C_{i i}}, \quad \mathrm{i}=1, \ldots, \mathrm{n} \tag{2.45}
\end{equation*}
$$

The quantity $t_{1-\alpha / 2}(n-r)$ is the $t$-statistic for $n-r$ degrees of freedom at the confidence level of $100(1-\alpha) \%$. For example, for $\mathrm{m}=20$ measurement, if $\mathrm{n}=2$ parameters are estimated, and if the $95 \%$ confidence is wanted, then $\alpha=0.05$ and $t_{0.975}(20-2)=2.1$. For a high number of measurements (>200), the $t$-statistic tends to the Gaussian statistic and we have $t_{0.975} \rightarrow 1.96$. Finally the result of the estimation process of the unknown exact parameter $x_{i}^{\text {exact }}$ can be presented in the following way:

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$$
\begin{gathered}
\text { ' } x_{i}^{\text {exact }} \text { has } 95 \% \text { chance of being in the interval }\left[\hat{x}_{i, \text { OLS }}-C l_{i}^{0.95}\right. \\
\text { or : ' } \left.x_{i}^{\text {exact }}=\hat{x}_{i, O L S} \pm C l_{i}^{0.95}+C I_{i}^{0.95}\right] \text { with } 95 \% \text { chance' } \\
\hline
\end{gathered}
$$

### 2.5.5. The residuals analysis

When estimation is achieved, the graphical analysis of the residuals given by eq. (2.13) can enable to detect some inconsistencies in the result. The difference between measurements and model response with optimal parameters must 'look like' the measurement noise, or in other words: 'the right model with the right parameters must explain the measurements except its random part'. For a noise with standard assumptions (independent and identically distributed, see end of section 2.5.2), the statistical properties of the residuals must be close to the measurement error properties (zero mean and variance $\sigma_{\varepsilon}^{2}$ ). If residuals are signed, the problem may be due to an error in the statistical assumptions regarding the measurements or in the structure or parameters of the direct model.


Figure 5 : graphical analysis of residuals at the end of the estimation

## 3. Indicators for a successful estimation

It has been shown that the matrix $\boldsymbol{s}^{t} \boldsymbol{S}$, also called the information matrix, is fundamental in the process of parameter estimation :

- it has to be inverted to achieve the OLS estimation according to (2.23)
- it also has to be inverted to compute the covariance matrix according to (2.35) associated to the OLS estimation. The diagonal terms of this matrix contain the variances of each estimation, and its off diagonal terms enable to compute the correlation matrix. The inverse of $\boldsymbol{s}^{t} \boldsymbol{s}$ play the role of "noise amplifier",
- the eigenvalues of $\boldsymbol{S}^{\boldsymbol{t} \boldsymbol{S}}$ enable the calculation of the lengths of the half principal axes of the elliptical confidence region,
- the determinant of $\boldsymbol{s}^{t} \boldsymbol{s}$ enables the calculation of the area of the elliptical confidence region.

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The difficulty is clear: $\boldsymbol{S}^{\boldsymbol{t} \boldsymbol{S}}$ has to be non-singular to be inverted and $\boldsymbol{s}^{\boldsymbol{t} \boldsymbol{S}}$ has to be not 'quasi-singular' in order to limit the noise amplification. This notion of non-singular character of the information matrix $\boldsymbol{S}^{t} \boldsymbol{S}$ makes sense only if all the parameters $x_{j}$ have the same physical units. Otherwise, one should study matrix $\boldsymbol{S}^{*} \boldsymbol{S}^{*}$ where $\boldsymbol{S}^{*}$ is the reduced sensitivity matrix, see section 3.1 and 3.3.

Then we have to find some indicators to evaluate the singularity and the quasi-singularity of $\boldsymbol{s}^{t} \boldsymbol{S}$. The first indication can be simply graphical. Singularity exists if a sensitivity coefficient $S_{i}(t)$ is purely proportional to another one $S_{i}(t)$; in that case the rank of $\boldsymbol{s}^{t} \boldsymbol{s}$ is lower than $n$, and its determinant is zero. More difficult is to find a linear combination of more than two sensitivity coefficients for which the consequences would be the same. The quasi-singularity would happen if sensitivity coefficients are 'similar' for all values of the independent variable (time here). This case happens most of the time: $\boldsymbol{s}^{\dagger} \boldsymbol{S}$ has full rank but its determinant is low and its condition number built with the ratio of its extreme eigenvalues:

$$
\begin{equation*}
\operatorname{cond}\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)=\frac{\lambda_{\max }\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)}{\lambda_{\min }\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)} \tag{2.46}
\end{equation*}
$$

takes high values.
'Visual' and 'quantitative' criteria will now be illustrated. We introduce first the reduced sensitivity matrix $\boldsymbol{S}^{*}$, that enables to compare the sensitivity coefficient between themselves and to compute a covariance matrix associated to relative estimations (and then to compute directly relative standard deviation associated to each parameter)

### 3.1. The reduced sensitivity matrix $\boldsymbol{S}^{*}$

It is given by

$$
\begin{gather*}
\boldsymbol{S}^{*}=\boldsymbol{S} \operatorname{diag}(\boldsymbol{x})  \tag{2.47}\\
\text { with } \operatorname{diag}(\boldsymbol{x})=\left(\begin{array}{ccc}
x_{1} & 0 & 0 \\
0 & \ddots & 0 \\
0 & 0 & x_{n}
\end{array}\right) \tag{2.48}
\end{gather*}
$$

It is built with the reduced sensitivity coefficients that are defined as

$$
\begin{equation*}
S_{k}^{*}(t, \boldsymbol{x})=x_{k} S_{k}(t, \boldsymbol{x})=\left.x_{k} \frac{\partial y_{m o}(t, \boldsymbol{x})}{\partial x_{k}}\right|_{t, x_{j} \text { for } j \not \equiv k}=\left.\frac{\partial y_{m o}(t, \boldsymbol{x})}{\frac{\partial x_{k}}{x_{k}}}\right|_{t, x_{j} \text { for } j \neq k} \tag{2.49}
\end{equation*}
$$

Equation (2.49) shows that the reduced sensitivity $S_{k}^{*}$ represents the absolute variation of model $\partial y_{m o}(t, \boldsymbol{x})$ due to a relative variation of parameter $\partial x_{k} / x_{k}$. These reduced sensitivity

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coefficients have then the same unit as both model output $y_{m o}$ and standard deviation $\sigma_{\varepsilon}$ of the measurement noise. If their magnitude is less than the magnitude of the standard deviation $\sigma_{\varepsilon}$ of the measurement noise $\sigma_{\varepsilon}$, it means that the influence of the considered parameter on the model response will not be measurable with a correct accuracy. Consequently, the estimation of this parameter through the use of experimental measurements, if possible, will be highly inaccurate. Rapid information may then be given by comparing the magnitude of each reduced sensitivity coefficient to the magnitude of the measurement noise, with respect to the independent variable (here time).

In the preceding example, we have then (with $n=2$ parameters):

$$
\boldsymbol{S}^{*}=\left[\begin{array}{cccc}
S_{1}^{*}\left(t_{1}\right) & S_{2}^{*}\left(t_{1}\right) & \ldots & S_{n}^{*}\left(t_{1}\right)  \tag{2.50}\\
\vdots & \vdots & \vdots & \\
S_{1}^{*}\left(t_{i}\right) & S_{2}^{*}\left(t_{i}\right) & \ldots & S_{n}^{*}\left(t_{i}\right) \\
\vdots & \vdots & \vdots \\
S_{1}^{*}\left(t_{m}\right) & S_{2}^{*}\left(t_{m}\right) & \ldots & S_{n}^{*}\left(t_{m}\right)
\end{array}\right]=\left[\begin{array}{ccc}
S_{1}^{*}\left(t_{1}\right) & S_{2}^{*}\left(t_{1}\right) \\
\vdots & \vdots \\
S_{1}^{*}\left(t_{i}\right) & S_{2}^{*}\left(t_{i}\right) \\
\vdots & \vdots \\
S_{1}^{*}\left(t_{m}\right) & S_{2}^{*}\left(t_{m}\right)
\end{array}\right]=\left[\begin{array}{cc}
x_{1} t_{1} & x_{2} \\
\vdots & \vdots \\
x_{1} t_{i} & x_{2} \\
\vdots & \vdots \\
x_{1} t_{m} & x_{2}
\end{array}\right]
$$

Let us notice that all the coefficients defining $\boldsymbol{x}$ have to be chosen in order to calculate (and compare) the reduced sensitivity coefficients: contrary to the sensitivity coefficients of a linear model, they do depend on the value of the parameter vector $\boldsymbol{x}$. That is why they are usually calculated using a nominal value for this vector, that is a value that is a priori expected to be close to its exact value in a parameter estimation problem.

### 3.1.1. Graphical analysis of reduced sensitivity coefficients

As said before, when nominal values of parameters have been chosen, it could be very instructive to plot all the reduced sensitivity coefficients composing each column of $\boldsymbol{S}^{*}$ on the same graph in order to 'visually' detect some future bad conditioning of matrices $\boldsymbol{S}^{* t} \boldsymbol{S}^{*}$ and $\boldsymbol{S}^{\boldsymbol{t}} \boldsymbol{S}$. This may be caused by (see Beck et al., 1977):

- one or more columns of $\boldsymbol{S}^{*}$ have coefficients with low absolute values compared to the other ones and compared to the noise level $\sigma_{\varepsilon}$, indicating poor sensitivities of the model to some parameters,
- two or more column are close to be linearly dependent, indicating correlations between some parameters that will prevent their simultaneous identification. The simplest detectable dependence is the proportionality between two coefficients (see Figure 6 and Figure 7 for favorable and unfavorable situations)

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(a)

(d)

(b)

(c)

©

(1)

(g)

(a)

(b)

(c)

(d)

(9)

(e)

(b)

(4)

(i)

(b)

(i)

Figure 6 : some situations where reduced sensitivity coefficients $S_{k}^{*}$ and $S_{j}^{*}$ are linearly independent

Figure 7 : some situations where reduced sensitivity coefficients $S^{*}{ }_{k}$ and $S_{j}^{*}$ (and sometimes $S_{p}^{*}$ ) are linearly dependent, implying a bad conditioning of the information matrix $S^{* t} S^{*}$ making it difficult, or impossible, to inverse it.

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Complementary analysis of the scaled (or reduced) sensitivity matrix can also be found in section 3.3.2 of Lecture 4 of this series.

### 3.1.2. The relative covariance matrix, and relative confidence intervals

The relative covariance matrix (size $n \times n$ for estimation of $n$ parameters) is built in the same way than the absolute covariance matrix (see equations (2.35) and (2.39)) but the amplification matrix (inverse of information matrix) is now built with the reduced sensitivity matrix $\boldsymbol{S}^{*}$ instead of $\boldsymbol{S}$ :

$$
\boldsymbol{C}^{*}=\left[\boldsymbol{S}^{* t} \boldsymbol{S}^{*}\right]^{-1} \sigma_{\varepsilon}^{2}=\left[\begin{array}{cccc}
\left(\frac{\sigma_{1}}{\hat{x}_{1, \text { OLS }}}\right)^{2} & \frac{\operatorname{cov}\left(e_{1}, e_{2}\right)}{\hat{x}_{1, \text { OLS }}^{2}} & \ldots & \left.\frac{\operatorname{cov}\left(e_{1}, e_{n}\right.}{\hat{x}_{1, \text { OLS }}^{2}}\right)  \tag{2.51a}\\
& \left(\frac{\sigma_{2}}{\hat{x}_{2, \text { OLS }}}\right)^{2} & & \frac{\operatorname{cov}\left(e_{2}, e_{n}\right)}{\hat{x}_{1, \text { LLS }}^{2}} \\
& & \ddots & \vdots \\
\text { sym } & & & \left(\frac{\sigma_{n}}{\hat{x}_{n, \text { OLS }}}\right)^{2}
\end{array}\right]
$$

Then, the $\boldsymbol{C}^{*}$ contains on its main diagonal the $n$ relative variances associated to each component of the estimated vector $\hat{\boldsymbol{x}}_{\text {OLS }}$. The square root of the $\mathrm{i}^{\text {th }}$ diagonal component of $\boldsymbol{C}^{*}$ is then the relative standard deviation (dimensionless) associated to the estimation $\hat{\boldsymbol{x}}_{i, O L S}$ and can be expressed in \%.

$$
\begin{equation*}
\sqrt{C_{i i}^{*}}(\%)=\frac{\sigma_{i}}{\hat{x}_{i, O L S}}, \quad \mathrm{i}=1, \ldots, \mathrm{n} \tag{2.51b}
\end{equation*}
$$

Last, the half width of relative confidence interval $\mathrm{Cl}_{i}^{1-\alpha}(\%)$, at a level of confidence of $100(1-\alpha) \%$, associated to the estimation $\hat{x}_{i, \text { OLS }}$ (and that was evaluated with 100 stochastic simulations in part 2.5.1.) is now given by :

$$
\begin{equation*}
C l_{i}^{1-\alpha}(\%)=t_{1-\alpha / 2}(n-r) \times \sqrt{C_{i i}^{*}}, \quad \mathrm{i}=1, \ldots, \mathrm{n} \tag{2.52}
\end{equation*}
$$

Finally the result of the estimation process of the unknown exact parameter $x_{i}^{\text {exact }}$ can be presented as the following, with the relative confidence interval :

$$
\begin{gathered}
\text { ' } x_{i}^{\text {exact }} \text { has } 95 \% \text { chance of being in the interval }\left[\hat{x}_{i, O L S}-C l_{i}^{0.95}(\%) \quad \hat{x}_{i, O L S}+C l_{i}^{0.95}(\%)\right] \text { ' } \\
\text { or : ' } x_{i}^{\text {exact }}=\hat{x}_{i, O L S} \pm C l_{i}^{0.95}(\%) \text { with } 95 \% \text { chance' }
\end{gathered}
$$

The elliptical relative confidence region corresponding to the scattering of estimations of Figure 3 can also be computed with the relative information matrix $\boldsymbol{S}^{*} \boldsymbol{S}^{*}$, the resulting equation expressed in the reduced coordinates $\delta x^{*}$ is :

$$
\begin{align*}
& \boldsymbol{\delta} \boldsymbol{x}^{* t} \cdot \boldsymbol{S}^{* t} \boldsymbol{S}^{*} \cdot \boldsymbol{\delta} \boldsymbol{x}^{*}=\Delta^{2}  \tag{2.53}\\
& \boldsymbol{\delta} \boldsymbol{x}^{*}=\boldsymbol{\delta} \boldsymbol{x} \cdot \operatorname{diag}\left(\hat{\boldsymbol{x}}_{O L S}\right)^{-1}
\end{align*}
$$

'Absolute' and 'relative' ellipses are plotted respectively on Figure 9 and Figure 9 to show that they correctly predict the extent of the 100 estimations cloud.


Figure 8 : 100 estimations cloud and 95\% absolute confidence elliptical region around the cloud mean


Figure 9 : 100 estimations cloud and 95\% relative confidence elliptical region

### 3.2. Illustration, with a simple example, of different situations that modify the quality of estimation

Here are illustrated, with the example described by Table 2, the influence of some experimental parameters on the quality of estimation. This quality is visualized by the extent of the confidence region and some of the quantitative indicators presented above are also observed.

### 3.2.1. Influence of noise standard deviation $\sigma_{\varepsilon}$

In Figure 10 the extension of the confidence region is shown, with respect to the standard deviation of noise measurement $\sigma_{\varepsilon}$, without changing its orientation. This is conform to eq. (2.44) giving the ellipse area proportional to the square of $\sigma_{\varepsilon}$.

### 3.2.2. Influence of number of measurements $m$ (in the same time range)

In Figure 11 the extension of the confidence region is shown, with respect to the number of measurements $m$, without changing its orientation. This is conform to eq. (2.44) giving the

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ellipse area inversely proportional to the square root of $\operatorname{det}\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)$, this area being inversely proportional to $m$. Then halving the noise level is better than doubling the number of measurements.


### 3.2.3. Influence of time range (for $m=20$ measurements)

The last tested experimental parameter is the time range, with a constant number of measurements ( $m=20$ ), see Figure 12. The results are presented in Figure 13 and Table 3.


Figure 12 : three time range are tested, giving three clouds of estimations on Figure 13

Figure 13 shows that when experiments are done at 'high' time values, the confidence ellipse is growing, especially along the $\mathrm{x}_{2}$ axis: the estimation of $\mathrm{x}_{2}$ (intercept of the model $x_{1} t+x_{2}$ ) is more and more inaccurate when the measurements are realized at high time values (far from $t=0$ ). This is confirmed by the reduced sensitivity plots on Figure 14 and Figure 15 (see comments in legends).

| Time range $(\mathrm{h})$ | $0.5 \mathrm{~h}-2.5 \mathrm{~h}$ | $5 \mathrm{~h}-7.5 \mathrm{~h}$ | $15 \mathrm{~h}-17.5 \mathrm{~h}$ |
| :--- | :--- | :--- | :--- |
| Central value $\bar{X}_{1}(\mathrm{~K} / \mathrm{h})$ | $4.994 \mathrm{~K} / \mathrm{h}$ | $4.738 \mathrm{~K} / \mathrm{h}$ | $4.985 \mathrm{~K} / \mathrm{h}$ |
| Absolute interval $(\mathrm{K} / \mathrm{h})$ | $\pm 0.3 \mathrm{~K} / \mathrm{h}$ | $\pm 0.35 \mathrm{~K} / \mathrm{h}$ | $\pm 0.35 \mathrm{~K} / \mathrm{h}$ |
| Relative interval $(\%)$ | $\pm 6 \%$ | $\pm 7 \%$ | $\pm 7 \%$ |
| Central value $\bar{X}_{2}(\mathrm{~K})$ | 2.019 | 3.52 | 2.223 |
| Absolute interval $(\mathrm{K})$ | $\pm 0.5 \mathrm{~K}$ | $\pm 1 \mathrm{~K} / \mathrm{h}$ | $\pm 5.3 \mathrm{~K} / \mathrm{h}$ |
| Relative interval $(\%)$ | $\pm 10 \%$ | $\pm 28 \%$ | $\pm 106 \%$ |

Table 3 : results of estimations for three different time ranges,withm=20 measurements

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Figure 14 : first time range (between 0.5h and 2.5h). Reduced sensitivity are of same order of magnitude, sensitivity to $x_{1}$ is better than $x_{2}$ and is increasing with time


Figure 15 : third time range (between $15 h$ and 17.5h). Reduced sensitivity to $x_{1}$ is far better than sensitivity to $x_{2}$ that appears now very close to zero comparing to $\mathrm{S}_{1}^{*}$

Last, the Table 4 shows multiple indicators confirming that increasing the beginning of the time range for the estimation of $x_{1}$ and $x_{2}$ is degrading the conditioning and then the quality of estimation.

| Time range $(\mathrm{h})$ | $0.5 \mathrm{~h}-2.5 \mathrm{~h}$ | $5 \mathrm{~h}-7.5 \mathrm{~h}$ | $15 \mathrm{~h}-17.5 \mathrm{~h}$ |
| :--- | :--- | :--- | :--- |
| $\lambda_{\text {min }}$ of $\boldsymbol{S}^{*} \boldsymbol{S}^{*} \uparrow$ | 1.03 e 1 | $6.5 \mathrm{e}-1 \downarrow$ | $6.2 \mathrm{e}-2 \downarrow$ |
| $\lambda_{\max }$ of $\boldsymbol{S}^{*} \boldsymbol{S}^{*} \downarrow$ | 1.29 e 3 | $1.8 \mathrm{e} 4 \uparrow$ | $1.3 \mathrm{e} 5 \uparrow$ |
| $\operatorname{det}\left(\boldsymbol{S}^{*} \boldsymbol{S}^{*}\right) \uparrow \uparrow$ | 1.34 e 4 | $1.18 \mathrm{e} 4 \downarrow$ | $8.0 \mathrm{e} 3 \downarrow$ |
| Ellipse area $\downarrow$ | $3.52 \mathrm{e}-4$ | $3.99 \mathrm{e}-4 \uparrow$ | $5.9 \mathrm{e}-4 \uparrow$ |
| cond $\left(\mathbf{S}^{*} \boldsymbol{S}^{*}\right)=\lambda_{\max } / \lambda_{\min } \downarrow$ | 1.24 e 2 | $2.78 \mathrm{e} 4 \uparrow \uparrow$ | $2.1 \mathrm{e} 6 \uparrow \uparrow$ |
| $\rho_{12} \downarrow$ | -0.93 | $-0.995 \uparrow$ | $-0.993 \leftrightarrow$ |

Table 4 : indicators values for the three experiments. In the first column, the arrows indicate if the indicator should be high (arrow up) or low (arrow down) to improve the conditioning.

### 3.3. Singular Value Decomposition of a matrix and condition number

### 3.3.1 Singular Value Decomposition of a rectangular matrix

Any rectangular matrix (called $\boldsymbol{K}$ here) with real coefficients and dimension ( $m, n$ ) with $m \geq n$, can be written under the form :

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This expression is sometimes called "lean" singular decomposition or "economical" SVD and involves

- U, an orthogonal matrix of dimensions $(m, n)$ : its column vectors (the left singular vectors of $\boldsymbol{K}$ ) have a unit norm and are orthogonal by pairs: $\boldsymbol{U}^{t} \boldsymbol{U}=\boldsymbol{I}_{n}$, where $\boldsymbol{I}_{n}$ is the identity matrix of dimension $n$. Its columns are composed of the first $n$ eigenvectors $\boldsymbol{U}_{k}$, ordered according to decreasing values of the eigenvalues of matrix $\boldsymbol{K} \boldsymbol{K}^{t}$. Let us note that, in the general case, $\boldsymbol{U} \boldsymbol{U}^{t} \neq \boldsymbol{I}_{m}$.
- $\boldsymbol{V}$, a square orthogonal matrix of dimensions $(n, n),: \boldsymbol{V} \boldsymbol{V}^{t}=\boldsymbol{V}^{t} \boldsymbol{V}=\boldsymbol{I}_{n}$. Its column vectors (the right singular vectors of $\boldsymbol{K}$ ), are the $n$ eigenvectors $\boldsymbol{V}_{\boldsymbol{k}}$, ordered according to decreasing eigenvalues, of matrix $\boldsymbol{K}^{t} \boldsymbol{K}$;
- $\boldsymbol{W}$, a square diagonal matrix of dimensions ( $n \times n$ ), that contains the $n$ so-called singular values of matrix $\boldsymbol{K}$, ordered according to decreasing values: $w_{1} \geq w_{2} \geq \cdots \geq w_{n}$. The singular values of matrix $\boldsymbol{K}$ are defined as the square roots of the eigenvalues of matrix $\boldsymbol{K}^{t} \boldsymbol{K}$. If matrix $\boldsymbol{K}$ is square and positive-definite, the eigenvalues and the singular values of $K$ are the same.

Another SVD form called "Full Singular Value Decomposition" is available for matrix $\boldsymbol{K}$. In this equivalent definition, both matrices $\boldsymbol{U}$ and $\boldsymbol{W}$ are changed: the matrix replacing $\boldsymbol{U}$ is now square (size $m \times m$ ) and the matrix replacing $\boldsymbol{W}$ is now diagonal but non square (size $m \times n$ ). In the case $m \geq n$, this can be written:
$\boldsymbol{K}=\boldsymbol{U}_{0} \boldsymbol{W}_{0} \boldsymbol{V}^{t}$ with $\boldsymbol{U}_{0}=\left[\begin{array}{ll}\boldsymbol{U} & \boldsymbol{U}_{\text {comp }}\end{array}\right] ; \boldsymbol{W}_{0}=\left[\begin{array}{c}\boldsymbol{W} \\ \mathbf{0}_{(m-n) \times n}\end{array}\right]$ and $\operatorname{dim}\left(\boldsymbol{U}_{\text {comp }}\right)=m \times(m-n)$
or:

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$$
[\boldsymbol{K}]=\left[\begin{array}{ll}
\boldsymbol{U} & \boldsymbol{U}_{\text {comp }}
\end{array}\right]\left[\begin{array}{ccc}
w_{1} & & \mathbf{0}  \tag{2.55b}\\
& \ddots & \\
\mathbf{0} & & w_{n} \\
0 & \cdots & 0 \\
\vdots & \ddots & \vdots \\
0 & \cdots & 0
\end{array}\right]\left[\begin{array}{ll} 
\\
\boldsymbol{V}^{t} \\
\end{array}\right]
$$

Matrix $\boldsymbol{U}_{\text {comp }}$ is composed of the $(m-n)$ left singular column vectors not present in $\boldsymbol{U}$. So, the concatenated matrix $\boldsymbol{U}_{0}$ verifies now:

$$
\begin{equation*}
\boldsymbol{U}_{0}^{t} \boldsymbol{U}_{0}=\boldsymbol{U}_{0} \boldsymbol{U}_{0}^{t}=\boldsymbol{U} \boldsymbol{U}^{t}+\boldsymbol{U}_{\text {comp }} \boldsymbol{U}_{\text {comp }}^{t}=\boldsymbol{I}_{m} \tag{2.56}
\end{equation*}
$$

This singular value decomposition (2.55b) can be implemented for any matrix $\boldsymbol{K}$, with real value coefficients, for $m \geq n$. Let us note that if, a decomposition of similar type exists too.

### 3.3.2 Interest of the Singular Value Decomposition in linear parameter estimation

We have seen above that if all the $n$ parameters in a parameter vector x are looked for for a linear model $\boldsymbol{y}_{m o}(\boldsymbol{x})=\boldsymbol{S} \boldsymbol{x}$, where $m$ noised measurements $\boldsymbol{y}=\boldsymbol{S} \boldsymbol{x}+\boldsymbol{\varepsilon}$ are available, and if noised $\boldsymbol{\varepsilon}$ is i.i.d., that is $\operatorname{cov}(\boldsymbol{\varepsilon})=\sigma_{\varepsilon}^{2} \boldsymbol{I}_{m}$, its OLS estimator can be written:

$$
\begin{equation*}
\hat{\boldsymbol{x}}_{O L S}=\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)^{-1} \boldsymbol{S}^{t} y \quad \text { with } \quad \mathrm{E}(\boldsymbol{\varepsilon})=\mathbf{0} \quad \text { and } \quad \operatorname{cov}\left(\hat{\boldsymbol{x}}_{O L S}\right)=\sigma_{\varepsilon}^{2}\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)^{-1} \tag{2.57}
\end{equation*}
$$

The potential difficulty in its estimation may stem from the possible ill-conditioning of the square information matrix $\boldsymbol{S}^{t} \boldsymbol{S}$ whose inversion make the standard deviations of its different parameters $\hat{x}_{j}$ become very large with respect to their exact value, see equation (2.35). So, a normalized critetion can be constructed in order to assess the quality of the estimation of the $n$ parameters.
This can be made through normalization of all the parameters $x_{j}$ present in parameter vector $\boldsymbol{x}$ by a a nominal value $\mathrm{x}_{\text {nom, } ;}$ (which, in parameter estimation results from a prior knowledge of the order of magnitude of the corresponding parameter) to get a reduced parameter vector $\boldsymbol{x}^{\text {red }}$ without any physical unit:

$$
\boldsymbol{x}^{\text {red }}=\boldsymbol{R}_{\text {nom }}^{-1} \boldsymbol{x}=\left[\begin{array}{c}
x_{1} / x_{1}^{\text {nom }}  \tag{2.58}\\
x_{2} / x_{2}^{\text {nom }} \\
\vdots \\
x_{n} / x_{n}^{\text {nom }}
\end{array}\right] \quad \text { with } \quad \boldsymbol{R}_{\text {nom }}=\left[\begin{array}{cccc}
x_{1}^{\text {nom }} & 0 & \cdots & 0 \\
0 & x_{2}^{\text {nom }} & \ddots & 0 \\
\vdots & \ddots & \ddots & \vdots \\
0 & 0 & \cdots & x_{n}^{\text {nom }}
\end{array}\right]
$$

So the output of the linear model can be expressed in terms of the reduced sensitivity matrix $\boldsymbol{S}^{\star}$ and of the reduced parameter $\boldsymbol{x}^{\text {red }}$ :

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$$
\begin{equation*}
\boldsymbol{y}_{m o}=\boldsymbol{S} \boldsymbol{x}=\boldsymbol{S} \boldsymbol{R}_{\text {nom }} \boldsymbol{R}_{\text {nom }}^{-1} \boldsymbol{x}=\boldsymbol{S}^{*} \boldsymbol{x}^{\text {red }} \quad \text { since } \boldsymbol{S}^{*}=\boldsymbol{S} \boldsymbol{R}_{\text {nom }} \tag{2.59}
\end{equation*}
$$

OLS estimation of this reduced parameter vector becomes, using equation (2.23):

$$
\begin{align*}
\hat{\boldsymbol{x}}^{\text {red }} & =\boldsymbol{R}_{\text {nom }}^{-1} \hat{\boldsymbol{x}}_{\text {OLS }}=\boldsymbol{R}_{\text {nom }}^{-1}\left(\boldsymbol{S}^{t} \boldsymbol{S}\right)^{-1} \boldsymbol{S}^{t} \boldsymbol{y}=\boldsymbol{R}_{\text {nom }}^{-1}\left(\boldsymbol{R}_{\text {nom }}^{-1} \boldsymbol{S}^{* t} \boldsymbol{S}^{*} \boldsymbol{R}_{\text {nom }}^{-1}\right)^{-1} \boldsymbol{R}_{\text {nom }}^{-1} \boldsymbol{S}^{* t} \boldsymbol{y}  \tag{2.60}\\
& =\boldsymbol{R}_{\text {nom }}^{-1} \boldsymbol{R}_{\text {nom }}\left(\boldsymbol{S}^{*} \boldsymbol{S}^{*}\right)^{-1} \boldsymbol{R}_{\text {nom }} \boldsymbol{R}_{\text {nom }}^{-1} \boldsymbol{S}^{* t} \boldsymbol{y}=\left(\boldsymbol{S}^{t t} \boldsymbol{S}^{*}\right)^{-1} \boldsymbol{S}^{* t} \boldsymbol{y}
\end{align*}
$$

And its covariance can be easily derived:

$$
\begin{equation*}
\operatorname{cov}\left(\hat{\boldsymbol{x}}^{\text {red }}\right)=\sigma_{\varepsilon}^{2}\left(\boldsymbol{S}^{* t} \boldsymbol{S}^{*}\right)^{-1} \tag{2.61}
\end{equation*}
$$

It is the same relationship as (2.51a). Since all the components of the reduced sensitivity matrix have the same unit as signal $\boldsymbol{y}$, and because $\boldsymbol{x}^{\text {red }}$ is dimensionless, it is possible to consider $\boldsymbol{S}^{*}$ as a linear application from a vector space of dimension $n$ into a vector space of dimension $m$. That was not possible for the original parameter column-vector $\boldsymbol{x}$, which did not belong to a true mathematical vector space, because its coefficients had not the same units.

So, it is now possible to write the lean SVD of $\boldsymbol{S}^{*}$, which uses the notion of Euclidian norm of different true vectors, see equation (2.54):

$$
\begin{equation*}
\boldsymbol{S}^{*}=\boldsymbol{U} \boldsymbol{W} \boldsymbol{V}^{t} \tag{2.62}
\end{equation*}
$$

It is now possible to calculate the amplification coefficient of the relative error $k_{r}$, see equation (1.7) in Lecture 1 of the same series:

$$
\begin{equation*}
k_{r}(\boldsymbol{\varepsilon})=\frac{\left\|\boldsymbol{e}_{\text {xred }}\right\| /\left\|\boldsymbol{x}_{\text {exact }}^{\text {red }}\right\|}{\|\boldsymbol{\varepsilon}\| /\left\|\boldsymbol{y}_{m o}\left(\boldsymbol{x}_{\text {exact }}^{\text {red }}\right)\right\|} \quad \text { with } \quad \boldsymbol{e}_{\text {xred }}=\hat{\boldsymbol{x}}^{\text {red }}-\boldsymbol{x}_{\text {exact }}^{\text {red }} \tag{2.63}
\end{equation*}
$$

Using the properties of matrices $\boldsymbol{U}$ and $\boldsymbol{V}$ described in section 3.3.1, as well as equation (2.60), one can show:

$$
\left.\begin{array}{l}
\left\|\boldsymbol{e}_{\text {xrea }}\right\|=\left\|\boldsymbol{V} \boldsymbol{W}^{-1} \boldsymbol{U}^{t} \boldsymbol{\varepsilon}\right\| \leq\left\|\boldsymbol{V} \boldsymbol{W}^{-1} \boldsymbol{U}^{t}\right\|\|\boldsymbol{\varepsilon}\|  \tag{2.64}\\
\left\|\boldsymbol{y}_{m o}\left(\boldsymbol{x}_{\text {exact }}^{\text {red }}\right)\right\|=\left\|\boldsymbol{S}^{*} \boldsymbol{x}^{\text {red }}\right\| \leq\left\|\boldsymbol{U} \boldsymbol{W} \boldsymbol{V}^{t}\right\|\left\|\boldsymbol{x}^{\text {red }}\right\|
\end{array}\right\} \Rightarrow k_{r}(\boldsymbol{\varepsilon}) \leq\left\|\boldsymbol{V} \boldsymbol{W}^{-1} \boldsymbol{U}^{t}\right\|\| \| \boldsymbol{U} \boldsymbol{W} \boldsymbol{V}^{t}\| \|
$$

One can recognize in the right-hand term of the last inequality (2.64) the product of norms of two matrices. The second matrix is simply the SVD form of the reduced sensitivity matrix $\boldsymbol{S}^{*}$ while the first one is just the pseudo inverse of $\boldsymbol{S}^{*}$, which is noted $\boldsymbol{S}^{*+}$ here.

Let's remind that the norm of any matrix $\boldsymbol{K}$ (which has not to be square) is defined by:

$$
\begin{equation*}
\|\boldsymbol{K}\|^{2}=\operatorname{Max}_{\|\boldsymbol{z}\|=1}\left(\boldsymbol{z}^{t} \boldsymbol{K}^{t} \boldsymbol{K} \boldsymbol{z}\right)=w_{1}^{2}(\boldsymbol{K}) \tag{2.65}
\end{equation*}
$$

where $\boldsymbol{w}_{1}(\boldsymbol{K})$ is the largest singular value of $\boldsymbol{K}$. This singular value is simply the square root of the largest (positive) value of the reduced information matrix $\lambda_{1}\left(\boldsymbol{S}^{*} \boldsymbol{S}^{*}\right)$, see equation (2.46). One can show that:

$$
\begin{equation*}
\left\|\boldsymbol{S}^{*}\right\|=w_{1}\left(\boldsymbol{S}^{*}\right) \text { and } \quad\left\|\boldsymbol{S}^{*}\right\|=w_{1}\left(\boldsymbol{S}^{*+}\right)=\frac{1}{w_{n}\left(\boldsymbol{S}^{*}\right)} \tag{2.66}
\end{equation*}
$$

So, it can be shown, using (2.63), (2.64) and (2.66) that the maximum value of the amplification coefficient of the relative error $k_{r}$, that is the criterion that assesses the ill-posed character of the OLS parameter estimation problem is equal to the condition number, noted cond (.) here, of the reduced sensitivity matrix:

$$
\begin{equation*}
k_{r}(\boldsymbol{\varepsilon}) \leq \operatorname{cond}\left(\boldsymbol{S}^{*}\right)=\frac{w_{1}\left(\boldsymbol{S}^{*}\right)}{w_{n}\left(\boldsymbol{S}^{*}\right)} \tag{2.63}
\end{equation*}
$$

So, this condition number, defined here with the Euclidian $L_{2}$ norm, is the pertinent criterion that can be used to measure the degree of ill-posedness of a linear parameter estimation problem, whatever the value of the noise level (for an i.i.d. noise). Since it requires the construction of the reduced sensitivity matrix, it depends on the nominal values of the parameters and can change strongly, depending on this choice, even if the problem is linear.

## 4. Conclusion

The example of a linear model with respect to its two parameters is rich enough to introduce many tools useful in the field of parameter estimation : sensitivity coefficients that compose the sensitivity matrix that has to be inverted to solve the estimation problem are the main tool. The covariance matrix that helps to qualify the quality of the estimation (variance of each estimation, correlation between them, size of the confidence region), needs also these coefficients. In the non linear case, the problem is often solved by assuming a local linear behaviour of the objective function to be minimized and of the confidence region.

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[^0]:    ${ }^{2}$ The objective of 'direct' modelisation is to give the best approximation of $\boldsymbol{y}_{\text {perfect }}$

[^1]:    ${ }^{3}$ it is here the more efficient, i.e. that will provide the estimation with the minimal variance

